

Review Paper.....

Various Techniques for Removal of Power Line Interference From ECG Signal

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Abstract— ECG is a biomedical signal which gives electrical activity of heart. This ECG signal is corrupted by various noises like power line interference, baseline wandering, channel noise, contact noise, muscle artifacts etc. Frequency range of ECG signal is nearly same as the frequency of power line interference. ECG signal has frequency range from 0.5 Hz to 80 Hz and power line interference introduces 50 to 60 Hz frequency component in that signal which is the major cause of corruption of ECG. This paper focuses on certain techniques for denoising of such non stationary ECG signal. Least mean square (LMS) algorithm effectively eliminate noise from signal but it requires reference model for analysis. So drawbacks of conventional method are overcome by wavelet transform hence wavelet transform is the best method for denoising the ECG signal.

Index Terms—ECG, FIR, IIR, PLI, EMF, FFT, STFT, DWT, IDWT

1 INTRODUCTION

The main function of the heart is to pump blood throughout the body to deliver the oxygen and nutrient demands of the body's tissues as well as to remove carbon dioxide. The first wave in ECG is called the P wave which is generated due to electrical activity of atria. Duration of atria should not be more than 0.11 seconds and amplitude should not be more than 3mm. The PR interval is measured from the beginning of the P wave to the beginning of the QRS complex. The normal duration for this is 0.12-0.20 seconds. The most important complex in the electrocardiogram is the QRS which shows electrical activities of ventricles. The duration (QRS interval) which is measured from the beginning of the QRS complex to its end is 0.05 to 0.10 seconds. The S-T segment follows the QRS complex. The duration of S-T segment is near about 0.08 seconds. Finally T wave represent repolarization of ventricles. The height of T wave should not be greater than 5 mm.

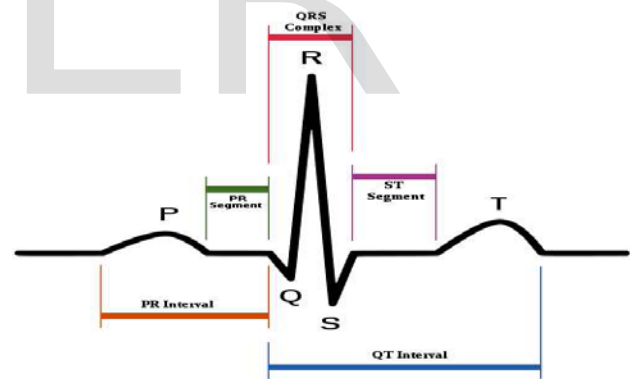


Fig.1 ECG Signal

This ECG signal is affected by various kind of noises like:

1. Power Line Interference:

The power line interference of 50/60 Hz is the source of interference and it corrupt the recordings of Electrocardiogram (ECG) which are extremely important for the diagnosis of patients. The interference is caused by:

- Electromagnetic interference by power line
- electromagnetic field (EMF) by the machinery which is placed nearby. The signal component holds harmonics with different amplitude and fre-

quency. The harmonics frequency is integral multiple of fundamental frequency such as 50Hz.

- c. Stray effect of the alternating current fields due to loops in the cables
- d. Improper grounding of ECG machine or the patient.
- e. Electrical equipments such as air conditioner, elevators and X-ray units draw heavy power line current, which induce 50 Hz signals in the input circuits of the ECG machine[2].

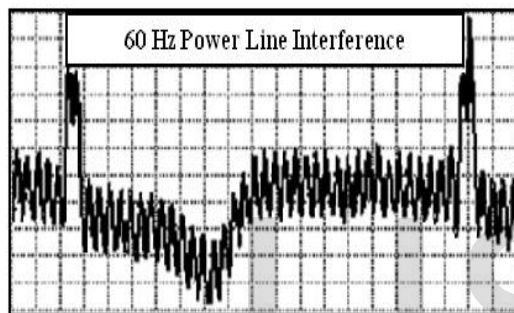


Fig.2 60 Hz Power Line Interference

The noise from electric power system is a major source of noise during the recording or monitoring of ECG. Different noises have different frequencies. The noise with low frequency is being problem with ECG signal as well as some time high frequency noises also interfere ECG like mobile phone. If the physical or mathematical variable changes rapidly then it can be high frequency and if it changes slowly then it would be low frequency. If the variable does not change at all then it is said that it has zero frequency. Most of the electronic devices such as ECG, transmitter, receiver, computer etc get power from power line. The 50 Hz alternative current (AC) is reduced in voltage, rectified and then filter to obtain low voltage direct current (DC). This is used to give power to those electronic devices.

2. Muscle Artifacts Known as an Electromyography (EMG) Noise:

This noise occurs at the time of muscle activity during an ECG recording especially in a stress test. This artifact consist of maximum frequency of 10 KHz

3. Data Collecting Device Noise:

This noise is mainly due to signal processing hardware

4. Patient-Electrode Motion Artifacts:

It is the movement of the electrode away from the contact area on the skin that leads to variations in the impedance between the electrode and skin causing potential variations in the ECG

5. Baseline Wandering :

Baseline wander is a low-frequency component present in the ECG system. This is due to offset voltages in the electrodes, respiration, and body movement. Baseline wander have frequency greater than 1Hz

6. Contact Noise:

This noise is caused by the loss of contact between the electrode and the skin, which effectively disconnects the measurement system and generates large artifacts since the ECG signal is usually capacitively coupled to the system. The characteristics of this noise signal include the amplitude of the initial transition, the amplitude of the 60 Hz component and the constant time of the decay.

7. Electrosurgical Noise:

Electrosurgical noise is generated by other medical equipment present in the patient care environment at frequencies between 100 KHz and 1 MHz .This noise remains approximately for 1 to 10 seconds.

8. Channel Noise:

Poor channel conditions can also introduce noise to the ECG when ECG is transmitted. It is mainly like white Gaussian noise which contains all frequency components.

Due to these interferences the quality of ECG signal can not be ideal so it is needed to improve the quality of required output of ECG signal.

2 Different Filtering Techniques Has Been Proposed For Cancellation of Power Line Interference From ECG Signal

2.1 IIR Notch Filter

Many of the researchers have used digital Infinite Impulse Response (IIR) filter to remove the effects of power line interference and baseline wander from ECG signals. Because, the design of IIR filter is simple. Notch filters can be used to remove the stationary power line interference. Stationary means which do not vary with amplitude, frequency and phase over time but ECG signal corrupted with power line interference is non stationary in nature and we never have exact prior knowledge of such PLI. In the absence of prior knowledge about power line interference (PLI) noise, application of notch or other conventional filters cause distortion of the ECG frequency spectrum [1]. For a second order notch filter, the bandwidth (Δf), notch frequency (f_0) and quality factor (referred as Q factor) are related by ,

$$Q = f_0 / \Delta f \quad (1)$$

for a fixed notch frequency, if Q is decreased, the bandwidth will increase and vice versa. The Q factor is decreased in order to increase attenuation level. If a notch filter has higher attenuation level, it will be able to remove PLI noise to a greater extent from ECG signal. The IIR notch filter practically fails to eliminate the line interference at frequencies other than 50 Hz,

Major drawbacks of IIR notch filter are:

1. With increase in attenuation level the PLI noise will removes effectively but It causes increase in notch filter bandwidth which eventually disturbs the nearby spectrum. This is the major drawback of using notch filter.
2. More filtering time is required
3. Memory requirement is also much more

4. Notch filter is incapable to filter the highly non-linear signals in the entire ECG range[1].

2.2 The FIR Filtering

The digital filters are divided into two basic types, Finite Impulse Response (FIR) and Infinite Impulse Response (IIR) filters. which are known as non recursive and recursive filters. FIR filter was chosen since it is simple and stable. The choice between the filter type (recursive and non recursive) is done due to the computational property and the storage required for the implementation.

The simplest method of FIR filter design is called the window method. A window is an array $w[n]$ consisting of coefficients that meet proposed filter requirements. The design of the FIR filter using the window method requires specifying which window functions is used. All frequencies below the cutoff frequency f_c are passed with unity amplitude while all higher frequencies are blocked. By taking the inverse Fourier Transform of this ideal frequency response, the ideal filter kernel (impulse response) is obtained.

The FIR filters are stable and having linear phase characteristics. FIR filters are having a transfer function of a polynomial in z-plane and is an all-zero filter that means the zeros in the z-plane determine the frequency response magnitude characteristic. The z transform of N-point FIR filter is given by[12]

$$H(z) = \sum_{n=0}^{N-1} h(n) z^{-n} \quad (2)$$

FIR filters are particularly useful for applications where exact linear phase response is required. The FIR filter is generally implemented in a non-recursive way which guarantees a stable filter. FIR filter using different windows are preferred due to ease of design and simplicity of programming .The various windows used are[12]:

Rectangular window:

$$W_R(n) = 1, 0 \leq n \leq M-1$$

$$= 0, \text{ Otherwise}$$
(3)

Hanning Window:

$$W_R(n) = 0.5 - 0.5 * \cos \left[\frac{2\pi n}{M-1} \right], 0 \leq n \leq M-1$$

$$= 0, \text{ Otherwise}$$
(4)

Hamming Window:

$$W_R(n) = 0.54 - 0.46 * \cos \left[\frac{2\pi n}{M-1} \right], 0 \leq n \leq M-1$$

$$= 0, \text{ Otherwise}$$
(5)

Blackman Window:

$$W_R(n) = 0.42 - 0.5 * \cos \left[\frac{2\pi n}{M-1} \right], 0 \leq n \leq M-1$$

$$= 0, \text{ Otherwise}$$
(6)

In rectangular window based FIR filter response, it was clear that the filter has sharp attenuation and pulsation in the stop band. In the pass band, the filter was found to be stable. The hamming, hanning and the blackman windows do not have a sharp cut-off like the rectangular window. Using these windows, we can designed the high pass filter of cut-off frequency 3 Hz and the low pass filter of cut-off frequency 100Hz. The noisy signals can passed through different filters to remove noise . Rectangular window based FIR filter can give good result as compare to other window [7].

The efficiency of different windows in case of FIR filter can be analyzed by evaluating following SNR[5],

$$SNR = 10 \log_{10} \frac{\sum(x_{denoised})^2}{\sum(x_{original} - x_{denoised})^2}$$
(7)

2.3 Adaptive Filter

An adaptive filter is a filter that self-adjusts its transfer function according to an optimization algorithm driven by an error signal. Because of the complexity of the optimization algorithms, most adaptive filters are digital filters. The adaptive filter reduces the mean squared error between primary input (ECG signal) and the reference input (noise with ECG signal). The power line interference (50Hz) from ECG signal can be removed by adaptive filtering while it's harmonics and high frequency noise can be removed by implementing general notch rejection filters.

A non-adaptive filter has a static transfer function while Adaptive filters can be used in applications where some parameters of the desired processing operation are not known in advance. The adaptive filter uses feedback in the form of an error signal to refine its transfer function to match the changing parameters. A filter can be used to remove the noise, extract information signals and separate two or more combined signals. if the a signal $x(k)$ is processed in a discrete system the output signal will be $y(k)$, if this output signal $y(k)$ is different from the original signal $x(k)$ then it must be needed to modify the system to get the required output. Then digital filter will be the solution to manipulate this problem. Digital filters are extremely used in noise cancellation, echo cancellation and also in the field of biomedical engineering to remove unwanted noise from ECG.

The word adaptive means to adjust with other environment (system) by having the same response as the system itself to some phenomenon which is taking place in it's surroundings. Or the system which tries to adjust its parameter, depending upon the other system's behavior and it's surrounding. The systems which carries out its functionality after undergoes the process of adaptation is called filter. The term 'filter' means to take the unnecessary particles (frequency component) from its input signal and process them to generate required output under certain specific rules. There are various principal option for the implemen-

tation of adaptive signal processing:

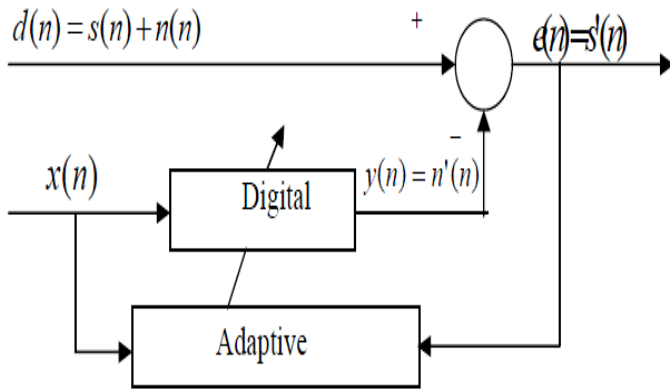


Fig.3 Adaptive Filter[16]

signal $d(n)$ contains not only desired signal $s(n)$ but also undesired noise signal $n(n)$. That is measured signal is distorted by noise $n(n)$. At that time, if undesired noise signal $n(n)$ is known, desired signal $s(n)$ can be obtained by subtracting noise signal $n(n)$ from corrupted signal $d(n)$. However complete information about noise source is difficult to obtain, thus estimated noise signal $n'(n)$ is used. The estimate noise signal $n'(n)$ is calculated through some filters and measurable noise source $X(n)$ which is linearly related with noise signal $n(n)$. After that, using estimated signal $n'(n)$ and obtained signal $d(n)$, estimated desired signal $s'(n)$ can be obtained. If estimated noise signal $n'(n)$ is more close to real noise signal $n(n)$, then more desired signal is obtained. Adaptive filter is classified into two parts, adaptive algorithm and digital filter. Function of adaptive algorithm is making proper filter coefficient. General digital filters use fixed coefficients, but adaptive filter change filter coefficients in consideration of input signal, environment, and output signal characteristics. Using this continuously changed filter coefficient, estimated noise signal $n'(n)$ is made by filtering $X(n)$ [16].

2.3.1 LMS Algorithm.

The input signal is a sum of desired signal $d(n)$ and interfering noise $v(n)$

$$X(n) = d(n) + v(n) \tag{8}$$

The variable filter has a Finite Impulse Response (FIR) structure. For such structures the impulse response is equal to the filter coefficients. The coefficients for a filter of order P are defined as

$$W_n = \{W_n(0), W_n(1), W_n(2), \dots, W_n(p)\}^T \tag{9}$$

The error signal or cost function is the difference between the desired and the estimated signal,

$$\epsilon(n) = d(n) - \hat{d}(n) \tag{10}$$

The variable filter estimates the desired signal by convolving the input signal with the impulse response. In vector notation this is expressed as

$$\hat{d}(n) = W_n * x(n) \tag{11}$$

where ,

$$X(n) = \{x(n), x(n-1), \dots, x(n-p)\}^T$$

is an input signal vector. The variable filter updates the filter coefficients at every time instant

$$W_{n+1} = w_n + \Delta w_n \tag{12}$$

where ΔW_n is a correction factor for the filter coefficients. The adaptive algorithm generates this correction factor based on the input and error signals.

Least mean squares (LMS) algorithms are a class of adaptive filter used to mimic a desired filter by finding the filter coefficients that relate to producing the least mean squares of the error signal (difference between the desired and the actual signal). It is a stochastic gradient descent method in that the filter is only adapted based on the error at the current time.

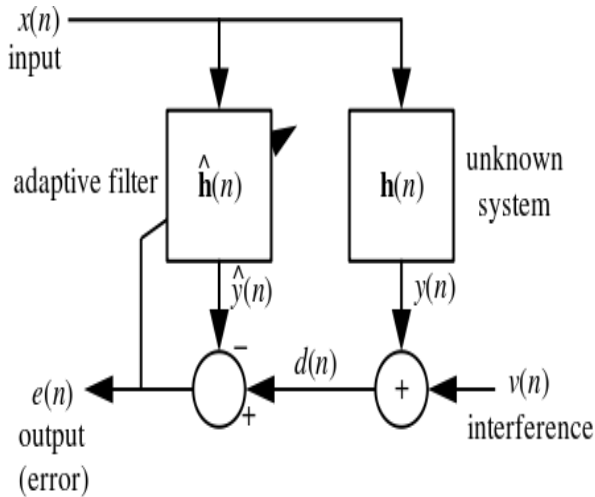


Fig.4 Least Mean Square Filter

An unknown system $h(n)$ is to be identified and the adaptive filter attempts to adapt the filter $\hat{h}(n)$ to make it as close as possible to $h(n)$. $x(n)$, $d(n)$ and $e(n)$ are observable signals while $y(n)$, $v(n)$, $h(n)$ are not directly observable.

Adaptive filtering methods effectively used for removing the power line interference and other noises from ECG Signals. Advantages of adaptive filter method are:

1. faster filtering response
2. smaller residual errors.

Drawback of adaptive filter method is:

1. This method requires reference signal (either signal or noise characteristics) information for the effective filtering process.

2.4 Wavelet Transform

In time domain analysis of ECG signal the analysis of additional or hidden Information is not possible. Hence transform from one domain to another is required. The transform of a function may give additional or hidden information about the original function, which may not be available otherwise. Thus Frequency domain representation of the function can be obtained by using fast fourier transform(FFT). Hence Perfect knowledge of what frequencies exists was obtained , but no information about where these frequencies are located in time.

Hence short time fourier transform come into existence to obtain the information of location of frequency in time. But again both time and frequency resolutions cannot be arbitrarily high. We cannot precisely know at what time instance a frequency component is located. We can only know what interval of frequencies are present in which time intervals. And in this STFT window size was fixed. wavelet analysis is superior to time domain analysis for identifying patients at increased risk of clinical deterioration. The transient nature of the ECG makes it ideal for WT analysis. WT allows a powerful analysis of non stationary signals, making it ideally suited for the high resolution interrogation of the ECG over a wide range of applications. The continuous wavelet transform (CWT) is a time–frequency analysis method which differs from the more traditional short time Fourier transform (STFT) by allowing arbitrarily high localization in time of high frequency signal features.

The CWT does this by having a variable window width, which is related to the scale of observation. Wavelet transform (WT) is designed to give good time resolution and poor frequency resolution at high frequencies, and good frequency resolution and poor time resolution at low frequencies. This approach is useful for ECG signals since ECG signals are characterized by high frequency components for short durations and low frequency components for long durations. WT allows a powerful analysis of non stationary signals, making it ideally suited for the high resolution interrogation of the ECG over a wide range of applications.

A continuous-time wavelet transform of $f(t)$ is defined as:

$$CWT \underline{x(t)} (\underline{a, b}) = w(\underline{a, b}) = \frac{1}{\sqrt{a}} \int_{-\infty}^{\infty} x(t) \cdot \Psi^* \left(\frac{t-b}{a} \right) dt$$

$$\text{Where } \Psi_{a,b}(t) = \frac{1}{\sqrt{a}} \Psi^* \left(\frac{t-b}{a} \right)$$

(13)

Here a , b belong to real number, $a \neq 0$ and they are dilating and translating coefficients, respectively. The asterisk de-

notes a complex conjugate. This multiplication of $\frac{1}{\sqrt{a}}$ is for the purpose of normalization of energy so that the transformed signal will have the same energy at every scale. The analysis function $\psi(t)$, the so-called mother wavelet, is scaled by a , so a wavelet analysis is often called a time-scale analysis rather than a time-frequency analysis. The wavelet transform decomposes the signal into different scales with different levels of resolution by dilating a single prototype function, the mother wavelet. Furthermore, a mother wavelet has to satisfy that it has a zero net area, which suggest that the transformation kernel of the wavelet transform is a compactly support function (localized in time), thereby offering the potential to capture the non-stationary spikes which normally occur in a short period of time.

with this choice of a and b , there exists the multiresolution analysis (MRA) algorithm, which decompose a signal into scales with different time and frequency resolution. MRA is designed to give good time resolution and poor frequency resolution at high frequencies and good frequency resolution and poor time resolution at low frequencies. The fundamental concept involved in MRA is to find the average features and the details of the signal via scalar products with scaling signals and wavelets. The wavelet decomposition results in levels of approximated and detailed coefficients. The time-frequency representation of DWT is performed by repeated filtering of the input signal with a pair of filters namely, low pass filter (LPF) and high pass filter (HPF), and its cutoff frequency is the middle of input signal frequency. The coefficient corresponding to the low pass filter is called as Approximation Coefficients and similarly, high pass filtered coefficients are called as Detailed Coefficients. The approximation coefficient is consequently divided into new approximation and detailed coefficients. This decomposition process is carried out until the required frequency response is achieved from the given input signal. This multi-resolution analysis enables us to analyze the signal in different frequency bands; therefore, we could observe any transient in time domain as well as in frequency

domain. The choice of mother wavelet can be selected based on correlation between the signal of interest and the wavelet-denoised signal. Discrete Wavelet Transform (DWT) based wavelet denoising have incorporated using different thresholding techniques to remove power line interference from ECG signal. Thresholding methods are used to denoise the ECG signals.

The general wavelet denosing procedure is as follows :

2.4.1 Apply wavelet transform to the noisy signal to produce the noisy wavelet coefficients to each level.

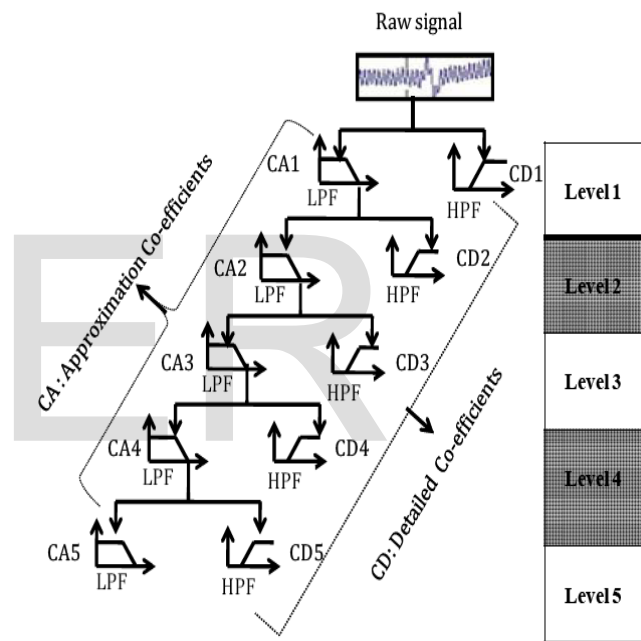


Fig.5 Filter bank structure for implementing DWT[14]

The time-frequency representation of DWT is performed by repeated filtering of the input signal with a pair of filters namely, low pass filter (LPF) and high pass filter (HPF), and its cutoff frequency is the middle of input signal frequency. The coefficient corresponding to the low pass filter is called as Approximation Coefficients (CA) and similarly, high pass filtered coefficients are called as Detailed Coefficients (CD) is shown in Figure. Furthermore, the CA is consequently divided into new approximation and detailed coefficients. This decomposition process is carried out until the required frequency response is achieved from the given input signal.

The relation between the low-pass and high-pass filter and the scalar function $\psi(t)$ and the wavelet $\varphi(t)$ can be states as following[17]

$$\varphi(t) = \sum_k h[k] \varphi[2t - k] \quad (14)$$

$$\psi(t) = \sum_k g[k] \varphi[2t - k] \quad (15)$$

The relation between the low-pass filter and high-pass filter is given as

$$g[L-1-n] = (-1)^n \cdot h[n] \quad (16)$$

where $g[n]$ is the high-pass, $h[n]$ is the low-pass filter, L is the filter length (total number of points). Filters satisfying this condition are commonly used in signal processing, and they are known as the Quadrature Mirror Filters (QMF).

At each level of decomposition the signal is decomposed into low and high frequencies. Due to the decomposition process the input signal must be a multiple of 2^n where n is the number of levels.

2.4.2 Down sampling of wavelet coefficient:

In DWT the decimation step removes every other of the coefficients of the current level. Thus the computation of the wavelet transform is faster and more compact in terms of storage space and one more important thing is, the transformed signal can be perfectly reconstructed from the remaining coefficients.

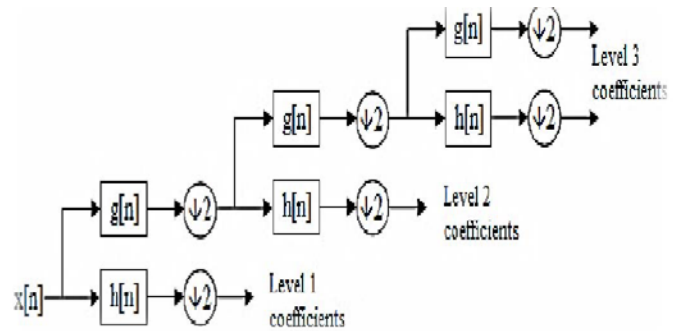


Fig.7 DWT Multilevel Decomposition

The two filtering and downsampling operation can be expressed by:

$$A^i[k] = \sum_n A^{i-1}(t) \cdot h[2k - n] \quad (17)$$

$$D^i[k] = \sum_n A^{i-1}(t) \cdot g[2k - n] \quad (18)$$

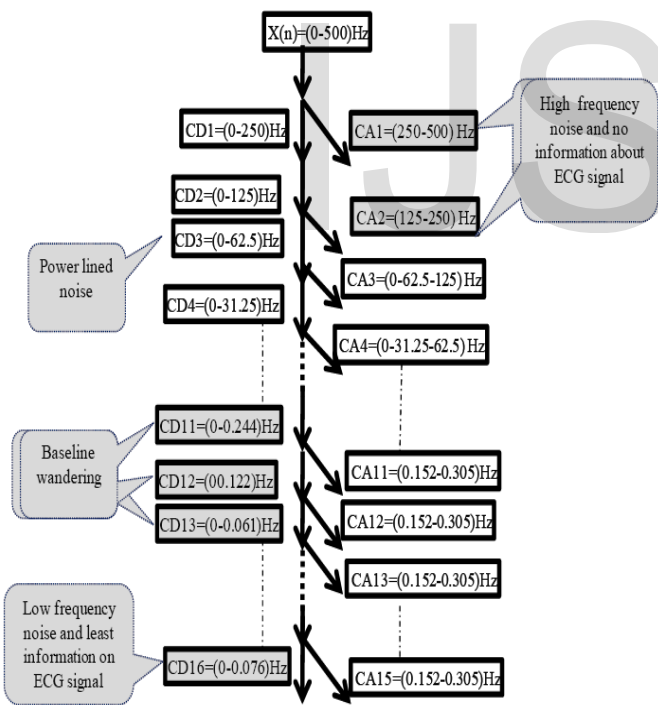


Fig.6 DWT Filter Structure With Relevant Noises[14]

This multi-resolution analysis enables us to analyze the signal in different frequency bands. Therefore we could observe any transient in time domain as well as in frequency domain.

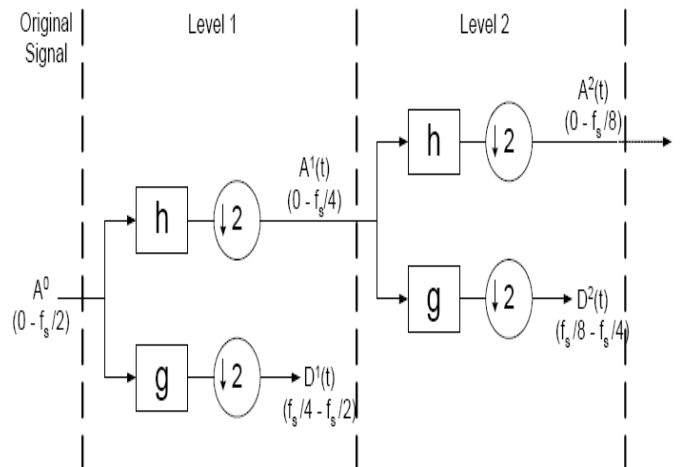


Fig.8 Multiresolution Wavelet Decomposition With Downsampling By 2 Operation[17]

Select appropriate threshold method and threshold rule at each level for quantization of wavelet coefficient.

cients to best remove the noises. Threshold value will be selected by obtaining a minimum error between wavelet coefficient of noise signal and original signal.

2.4.3 Various threshold methods available are as follows:

a. Hard Thresholding:

Hard thresholding method zeros the coefficients that are smaller than the threshold and leaves the other ones unchanged. Mathematically hard thresholding is given as[15],

$$D_H^T(w) = \begin{cases} w & \text{for all } |w| > T \\ 0 & \text{otherwise} \end{cases} \quad (19)$$

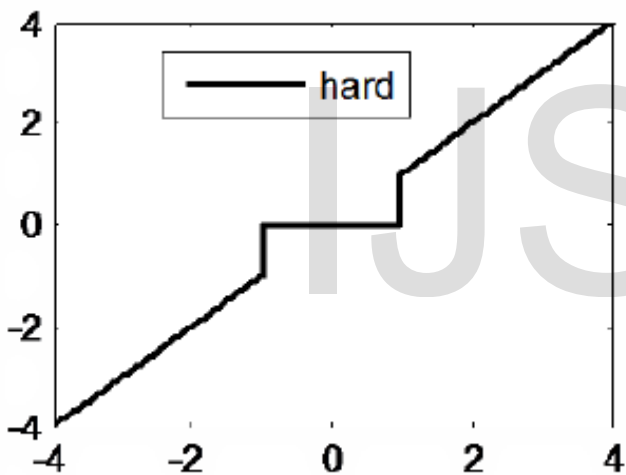


Fig.9 Hard Thresholding(Shrinkage) Wavelet Function[15]

b. The Soft Thresholding:

The soft thresholding method zeros the coefficient that are smaller than threshold and scales the remaining coefficients in order to form a continuous distribution of the coefficients centered on zero. Mathematically soft thresholding is given as[15]:

$$D_S^T(w) = \text{sgn}(w) \max(0, |w| - T) \quad (20)$$

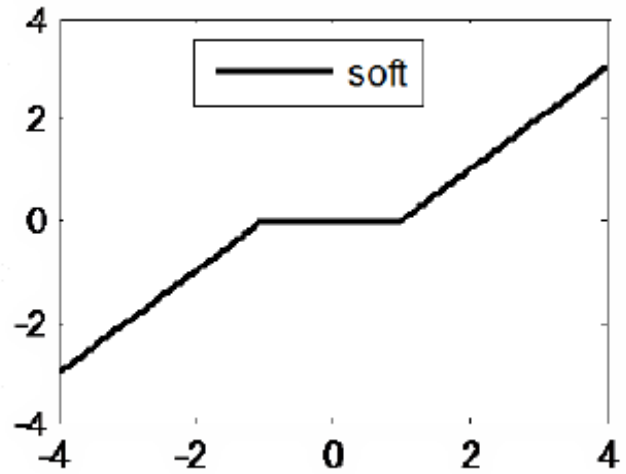


Fig.10 Soft Thresholding(shrinkage) Wavelet Function[15]

c. Semi Soft Thresholding:

Semi soft thresholding is a family of non-linearity that interpolates between soft and hard thresholding. It uses both a main threshold T and a secondary threshold $T_1 = \mu T$. When $\mu = 1$, the semi-soft thresholding performs a hard thresholding, whereas when $\mu = \infty$, it performs a soft thresholding. Mathematically semi soft thresholding is given as[15]:

$$D_{SS}^{T, T_1}(w) = \begin{cases} 0 & |w| \leq T \\ \text{sgn}(w) \frac{T_1(|w| - T)}{T_1 - T} & T < |w| \leq T_1 \\ w & |w| > T_1 \end{cases} \quad (21)$$

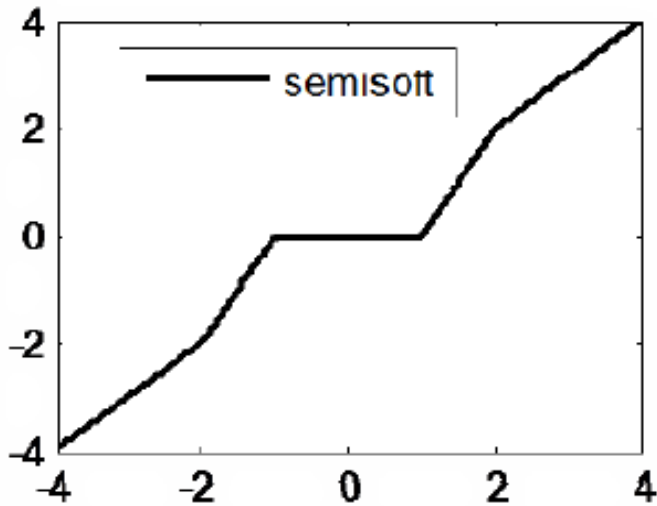


Fig.11 Semi Soft Thresholding(Shrinkage) Wavelet Function[15]

d. Stein Thresholding:

One more way to achieve a trade-off between hard and soft thresholding is to use a soft-squared thresholding non-linearity, this thresholding is also called as Stein estimator[15].

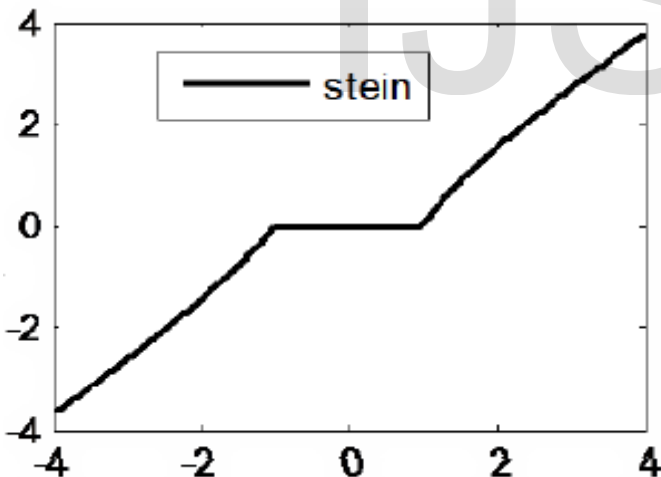


Fig.12 Stein Thresholding(Shrinkage) Wavelet Function[15]

2.4.4 Various thresholding rules are as follows:

a. Global Thresholding (w_{t_g}): This is a fixed threshold or global thresholding method and it is computed as[2]:

$$w_{t_g} = \sqrt{2 \log(n)} \quad (22)$$

where n is the total number of wavelet coefficients. This method yields the minimax performance is multiplied by the log value of the length of the wavelet coefficients.

b. Rigerous SURE Thresholding ($w_{t_{su}}$): Steins unbiased risk estimator (SURE) or rigrsure is an adaptive thresholding method which is based on Stein's unbiased likelihood estimation principle. This method computes is likelihood estimation first using the given threshold t , and then minimize the non-likelihood t , so the threshold has been obtained[2].

c. Heuristic SURE Thresholding ($w_{t_{th}}$): Hensure threshold is a combination of SURE and global thresholding method. If the signal-to noise ratio of the signal is very small, then the SURE method estimation will have more amounts of noises. In this type of situation, the fixed form threshold is selected by means of global thresholding method. Minimax threshold is also used fixed threshold and it yields minimax performance for Mean Square Error (MSE) against an ideal procedures. In Hensure method the threshold value will be selected by obtaining a minimum error between wavelet coefficient of noise signal and original signal[2].

d. MINMAX Thresholding: It uses a fixed threshold chosen to yield minimax performance for mean square error against an ideal procedure. The minimax principle is used in statistics in order to design estimators. Since the denoised signal can be assimilated to the estimator of the unknown regression function, the minimax estimator is the one that realizes the minimum of the maximum mean square error obtained for the worst function in a given set[2].

2.4.5 Inverse wavelet transform of the thresholded wavelet coefficients to obtain a denoised signal. The denoised signals reconstructed without affecting any features of signal of interest. More level of decomposition has to carried out in order to remove low frequency noise from signal. For

each level of decomposition threshold value has to calculate by applying threshold selection rule and the wavelet coefficient above the value of threshold has been removed (soft thresholding). The selection of hard and soft thresholding depends on application.

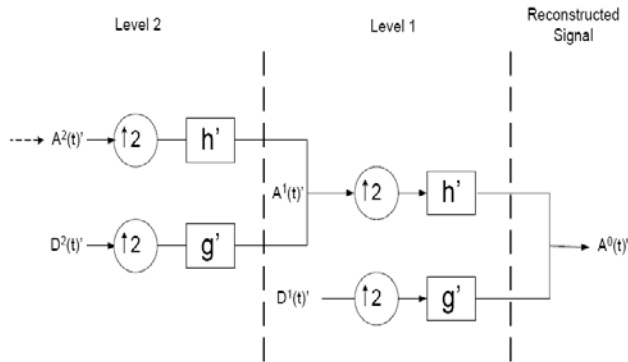


Fig.13 Multiresolution Wavelet Reconstruction Up Sampling by 2[17]

wavelet coefficient corresponding to frequency range of 50 to 60 HZ has to remove as it corresponds to power line interference. After applying threshold on each level of the original signal, the effects of noises on the ECG signals will get remove. Finally, we have to reconstruct the signal on each level by using Inverse Discrete Wavelet Transform (IDWT) to obtain noise free ECG signal.

3 CONCLUSION

Finite impulse response consist of limited number of impulses hence number of computation elements are also less hence this method is flexible and cost effective but these types of filters are generally realized non recursively or non adaptive which means that there is no feedback involved in computation of the output data. The output of the filter depends only on the present inputs. So as compare to least mean square and wavelet transform, FIR filter using various windows has less efficiency to remove 50 Hz power line interference from ECG signal. Thus we can say that among all the methods discussed above wavelet transform is the best method for denoising of ECG as it does not re-

quire any reference model and accuracy is also much better.

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